Research Article

Optimal State Estimation for Discrete-Time Markov Jump Systems with Missing Observations

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This paper is concerned with the optimal linear estimation for a class of direct-time Markov jump systems with missing observations. An observer-based approach of fault detection and isolation (FDI) is investigated as a detection mechanic of fault case. For systems with known information, a conditional prediction of observations is applied and fault observations are replaced and isolated; then, an FDI linear minimum mean square error estimation (LMMSE) can be developed by comprehensive utilizing of the correct information offered by systems. A recursive equation of filtering based on the geometric arguments can be obtained. Meanwhile, a stability of the state estimator will be guaranteed under appropriate assumption.

1. Introduction

Discrete-time Markov jump linear systems (MJLSs) are basically linear discrete-time systems with discretional parameters evolving with a finite-state Markov chain. It can be used in modeling systems with abrupt structures, for example, those which may be found in signal processing, fault detection [1, 2], and subsystem switching. One classical application is maneuvering target tracking, in which signals of interest are modeled by using MJLSs [3]. In these fields, the problems of state estimation for MJLSs play an essential role in recovering some desired variables from given noisy observations for output variables. However, many approaches of achieving the state estimation of MJLSs include the generalized pseudo-Bayesian (GPB) algorithm [4, 5], the interacting multiple model (IMM) filtering [6], stochastic sampling based methods [7, 8], and LMMSE filter. Those methods are different from each other in their estimation criteria and means [2, 9-12]. Among them, LMMSE filter has been well studied for MJLSs in many of literary works [9].

On the other hand, since applications of sensors networks are becoming ubiquitous in practical systems, wireless or wireline communication channels are essential for data communication. Examples are offered ranging from advanced aircraft, spacecraft to manufacturing process. As communication channels are time varying and unreliable, the phenomena of random time delays and random packet dropout usually occur in these networked systems. Hence, more and more attention has been paid to systems with observer-based fault during the past years. For example, studies on optimal recursive filter for systems with intermittent observations can be traced back to Nahi [13], whose work assumed that uncertainty of observations is independent and identically distributed. Afterwards, by using linear matrix inequalities (LMIs) techniques, the L_2 - L_∞ performance, H_∞ performance, finite-time L_2 - L_{∞} performance, and finitetime H_{∞} performance have been well studied for solving filtering and control problems occurring in stochastic systems with uncertain elements [14-21]. In [22-25], the stability analysis of random Riccati equation arising from Kalman filtering with intermittent observations was investigated elaborately. H_{∞} filtering algorithm [26-28] has been developed for discrete systems with random packet losses in [29, 30]. In [31], a robust filtering algorithm was developed for state estimation of MJLSs with random missing observation by applying basic IMM approach and H_{∞} technique. Reference [32] dealt with the fault detection filtering (FDF) design within stochastic H_{∞} filtering frame for a class of discrete-time nonlinear Markov jump systems with lost measurements.

Although the aforementioned references give efficient and practical tools to deal with the filtering problems for systems with package dropout, the results given by the methods constructed based on LMIs techniques are sometimes too conservative. What is more, IMM approach mentioned a priori requires online calculations. Inspired by the effectiveness of LMMSE mechanic used in solving state estimation problem of MJLSs with random time delays in [33], the problem of state estimation of MJLSs with random missing observations is formulated into LMMSE filtering frame. This frame can lead to a time-varying linear filter easy to implement. At the same time, most calculations can be performed off-line.

Aiming at solving the issue of uncertain observations in MJLSs, this paper provides a heuristic method for detecting the fault in process of transmitting observation. An approach of fault detection and isolation (FDI) [32, 34] for a class of MJLSs with missing observations will be investigated. The key point of FDI is to construct the residual generator and determine the residual evaluation function and the threshold. Then, by comparing the value of the evaluation function with the prescribed threshold, we will make judgment whether an alarm of fault is generated. The situation of uncertainties of observation can be naturally and conveniently reflected. With knowing the information of the faulty case, a conditional prediction of observation will be obtained, which can be used as replacement of the faulty one. At this time, we can utilize the optimal state estimator of pervious instant and parameters for constructing observer of system to estimate the observation at current time. By this way, we can skip and avoid the fault observation.

Accordingly, by applying the basic FDI approach and basic LMMSE algorithm, an FDI-LMMSE filtering algorithm is developed for state estimation of MJLSs with random missing observation. In order to solve the optimal estimation problem, the measurements' loss process is modeled as a Bernoulli distributed white sequence taking values from 0 to 1 randomly. The estimation problem is then reformulated as an optimal linear filtering of a class of MJLSs, which have random missing observation and necessary model compensation, via state augmentation [35–38]. A recursive filtering is formulated in terms of Riccati difference equations. At the same time, we will show that estimator is stable under necessary assumptions in this paper.

This paper is organized as follows. Section 2 gives the problem formulation. A recursive optimal solution is given in Section 3. Its stability is discussed in Section 4. In Section 5, a numerical example is shown to explain the effectiveness of approach proposed in our paper. At last, the conclusions are drawn in Section 6.

2. Problem Formulation

On the stochastic basis $(\Omega, \mathcal{F}_k, \{\mathcal{F}_k\}, P)$, considering the following jump Markov linear system model:

$$x_{k+1} = A(r_k) x_k + B(r_k) (a(r_k) + w_k)$$

$$y_k = \gamma_k C(r_k) x_k + D(r_k) v_k,$$
(1)

where $\{x_k \in \mathbb{R}^n\}$ is continuous-valued based-state sequence with known initial distribution $x_0 = \mathcal{N}(x_0; \bar{x}_0, \Sigma_0)$. $a(r_k) \in$ R^n is assumed to be known time-varying constant to each value of r_k . $\{y_k \in R^s\}$ is the noisy observation sequence. $\{w_k \in R^n\}$ is the noisy observation sequence with distribution $w_k \sim N(w_k; 0, Q)$. $\{v_k \in R^s\}$ is a white measurement noise sequence independent of the process noise with distribution $v_k \sim N(v_k; 0, R)$.

Remark 1. $a(r_k)$ is a compensation between practical systems and models applied in this paper.

 $\{r_k\}$ is the unknown discrete-valued Markov chain with a finite-state space $N = \{1, 2, ..., N\}$. The transition probability matrix is $\Pi = [\pi_{ij}]_{N \times N}$, where $i, j \in N$. We set $\mu_i(k) := P(r_k = i)$. The basic variables w_k, v_k, x_0 and the modal-state sequence r_k are assumed to be mutually independent for all k. $A(r_k)$, $B(r_k)$, $C(r_k)$, $D(r_k)$ are assumed to be known time-varying system matrices to each value of r_k . For notational simplicity, the following notations and definitions hold in the rest of the paper:

$$r_k^j = \{r_k = j\}, \quad A_j = A\left(r_k^j\right), \quad B_j = B\left(r_k^j\right),$$

$$C_j = C\left(r_k^j\right), \quad D_j = D\left(r_k^j\right).$$
(2)

In this paper, consider that the observations are sent to the estimator via a Gilbert-Elliot channel, where the packet arrival is modeled using a binary random variable $\{\gamma_k\}$, with probability $P(\gamma_k = 1) = \eta$, and with γ_k independent of γ_s if $k \neq s$. Let γ_k be independent of w_k , v_k , x_k ; that is, according to this model, the measurement equation consists of noise alone or noise plus signal, depending on whether γ_k is 0 or 1.

Notation 1. Some notations which we will use throughout the paper should be presented first. We will denote by $\mathbb{R}^{m \times n}$ the space of $m \times n$ real matrices and by \mathbb{R}^m the space of mdimensional real vectors. The superscript T indicates transpose of a matrix. For a collection of N matrices D_1, \ldots, D_N , with $D_j \in \mathbb{R}^{m \times n}$, diag $\{D_j\} \in \mathbb{R}^{Nm \times Nn}$ represents the diagonal matrix formed by D_j in the diagonal.

Notation 2. Define $H^n = \{X = (X_1 \cdots X_N); X_i \in \mathbb{R}^{n \times n}, i \in N\}$ and $H^{+n} = \{X = (X_1 \cdots X_N); X_i \ge 0, i \in N\}$. For $X = (X_1 \cdots X_N) \in H^{+n}, V = (V_1 \cdots V_N) \in H^{+n}$, if $X \ge V$ for each $i \in N$, we have $X_i \ge V_i$.

3. Recursive Optimal Solution

In this section, a solution to the optimal estimator will be presented via the projection theory and the state augmentation in the Hilbert space.

3.1. Preliminaries. First, we denote by $\mathscr{L}(y^k)$ the linear space spanned by the observation $y^k = \{y_k^T, \dots, y_0^T\}$. If $\theta = \sum_{i=1}^k \xi(i)^T y_i$ for some $\xi(i) \in \mathbb{R}^m$, $i = 1, \dots, k$, the random variable $\theta \in \mathscr{L}(y^k)$.

Let $1_{\{r_k=j\}}$ represent an indicator of Markov process, which is defined as follows:

$$z_j(k) \triangleq x_k \mathbf{1}_{\{r_k=j\}} \in \mathbb{R}^n \tag{3}$$

$$z(k) \triangleq \begin{pmatrix} z_1(k) \\ \vdots \\ z_N(k) \end{pmatrix} \in \mathbb{R}^{Nn}.$$
(4)

And call $\hat{z}(k) = E(z(k))$. Define also $\hat{z}(k \mid k-1)$ as the projection of z(k) onto the linear space $\mathscr{L}(y^k)$ and

$$\widetilde{z}(k \mid k-1) \triangleq z(k) - \widehat{z}(k \mid k-1).$$
(5)

Then, we first define the following second-moment matrices associated with the aforementioned variables. They play key roles in deriving the covariance matrices of the estimator errors and optimal estimator:

$$Z_{i}(k) \triangleq E\left\{z_{i}(k)\left(z_{i}(k)\right)^{T}\right\} \in \mathbf{B}\left(\mathbb{R}^{n}\right),$$

$$Z(k) \triangleq E\left\{z\left(k\right)\left(z\left(k\right)\right)^{T}\right\} \in \mathbf{B}\left(\mathbb{R}^{Nn}\right),$$

$$\widehat{Z}(k \mid l) \triangleq E\left\{\widehat{z}\left(k \mid l\right)\left(\widehat{z}(k \mid l)\right)^{T}\right\} \in \mathbf{B}\left(\mathbb{R}^{Nn}\right),$$

$$\widetilde{Z}(k \mid l) \triangleq E\left\{\widetilde{z}\left(k \mid l\right)\left(\widetilde{z}(k \mid l)\right)^{T}\right\} \in \mathbf{B}\left(\mathbb{R}^{Nn}\right).$$
(6)

Considering the following augment matrices:

$$A(k) \triangleq \begin{bmatrix} \pi_{11}(k) A_{1}(k) \cdots \pi_{1N}(k) A_{N}(k) \\ \vdots & \ddots & \vdots \\ \pi_{N1}(k) A_{1}(k) \cdots \pi_{NN}(k) A_{N}(k) \end{bmatrix}$$

$$\in \mathbf{B}(R^{Nn})$$

$$D(k) \triangleq \begin{bmatrix} D_{1}(k) \mu_{1}(k)^{1/2} \cdots D_{N}(k) \mu_{N}(k)^{1/2} \end{bmatrix}$$

$$\in \mathbf{B}(R^{Ns}, R^{s}), \qquad (7)$$

$$C(k) \triangleq \begin{bmatrix} C_{1} \cdots C_{N} \end{bmatrix} \in \mathbf{B}(R^{Nn}, R^{s}), \qquad \overline{a} \triangleq \begin{bmatrix} a_{1}, \dots, a_{N} \end{bmatrix},$$

$$B(k) \triangleq \operatorname{diag}\left[\begin{bmatrix} (\pi_{k}^{1j} \mu_{k}^{1})^{1/2} B_{1} \cdots (\pi_{k}^{Nj} \mu_{k}^{N})^{1/2} B_{N} \end{bmatrix} \right]$$

$$\in \mathbf{B}(R^{N^{2}n}, R^{Nn})$$

then system can be described as follows:

$$z (k + 1) = A (k) z (k) + B (k) (\overline{a} + w (k))$$

$$y (k) = \gamma_k C (k) z (k) + D (k) v (k).$$
(8)

Note that $y(k) = y_k$.

Assumption 2. For all k, $B(k)\overline{a}\overline{a}^T B(k)^T \gg P$, where P convergency value of $\widetilde{Z}(k \mid k-1)$, which will be given in Section 3.

3.2. Optimal Estimator. From geometric arguments in [39], the LMMSE filter for MJLSs with uncertain observations can be derived in this section. The following lemmas present necessary and sufficient conditions on derivation of FDI-LMMSE filtering.

Lemma 3. For any given time instant k, one has

$$Z_{j}(k+1) = \sum_{i=1}^{m} \pi_{ij} A_{i}(k) Z_{i}(k) A_{i}^{T}(k) + \sum_{i=1}^{N} \pi_{ij} \mu_{k-1}^{i} B_{i}(a_{i}a_{i}^{T} + Q) B_{i}^{T}$$
(9)
$$Z(k) = \operatorname{diag} \left[Z_{j}(k) \right],$$

where $Z_i(0) = \mu_0^i X_0$.

Proof. For any given instant k, we have from (8) that

$$Z_{j}(k+1) = E\left[z_{j}(k+1)z_{j}^{T}(k+1)\right]$$
$$= \sum_{i=1}^{m} \pi_{ij}A_{i}(k)Z_{i}(k)A_{i}^{T}(k)$$
$$+ \sum_{i=1}^{N} \pi_{ij}\mu_{k-1}^{i}B_{i}\left(a_{i}a_{i}^{T}+Q\right)B_{i}^{T}.$$
(10)

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Recalling that $X_0 = E[\bar{x}_0 \bar{x}_0^T]$, initial covariance matrix $Z_i(0) = \mu_0^i X_0$.

To derive the optimal filter, we first define the innovation sequence as

$$\widetilde{y}(k) = y(k) - \widehat{y}(k \mid k - 1), \qquad (11)$$

where conditional prediction $\hat{y}(k \mid k-1)$ is the projection of y(k) onto the linear space of $\mathcal{L}(y^{k-1})$. Consider

$$\hat{y}(k \mid k-1) = C(k)\hat{z}(k \mid k-1).$$
 (12)

Then, according to (4) and (8), the generated residual will be obtained as

$$\widetilde{y}(k \mid k-1) = \begin{cases} C(k)\widetilde{z}(k \mid k-1) + D(k)v(k), & \gamma_k = 1, \\ D(k)v(k) - C(k)\widehat{z}(k \mid k-1), & \gamma_k = 0. \end{cases}$$
(13)

In the following, an FDI scheme will be constructed, which can detect whether observation at instant k is lost. In this paper, we choose the following mean square of the residual as the residual evaluation function to measure the energy of the residual:

$$S_{k} = E\left(\widetilde{y}\left(k \mid k-1\right)\widetilde{y}\left(k \mid k-1\right)^{T}\right).$$
(14)

From (12)-(13) we get that

$$E\left(\tilde{y}(k \mid k-1) \ \tilde{y}(k \mid k-1)^{T}\right) = \begin{cases} C(k) \ \tilde{Z}(k \mid k-1) C(k)^{T} + D(k) \ RD(k)^{T}, & \gamma_{k} = 1, \\ C(k) \ \hat{Z}(k \mid k-1) C(k)^{T} + D(k) \ RD(k)^{T}, & \gamma_{k} = 0. \end{cases}$$
(15)

Suppose that $\widetilde{Z}(k \mid k-1)$ is convergent to *P* at the instant *k*, from Assumption 2, we have that

$$B(k-1)\overline{a}\overline{a}^{T}B(k-1)^{T} \gg \widetilde{Z}(k \mid k-1).$$
(16)

If $\gamma_k = 1$, we have that

$$S_{k}^{\gamma_{k}=1} = C(k) \widetilde{Z}(k \mid k-1) C(k)^{T} + D(k) RD(k)^{T}$$

$$\ll C(k) B(k-1) \overline{a} \overline{a}^{T} B(k-1)^{T} C(k)^{T} + D(k) RD(k)^{T}.$$
(17)

If $\gamma_k = 0$,

$$S_{k}^{\gamma_{k}=0} = C(k) \,\widehat{Z}(k \mid k-1) \,C(k)^{T} + D(k) \,RD(k)^{T}.$$
(18)

From (8),

$$\widehat{Z}(k \mid k-1) = A(k-1)\widehat{Z}(k-1 \mid k-1)A(k-1)^{T} + B(k-1)\overline{a}\,\overline{a}^{T}B(k-1)^{T}$$
(19)
$$\geq B(k-1)\overline{a}\,\overline{a}^{T}B(k-1)^{T};$$

we have that

$$S_{k}^{\gamma_{k}=0} \ge C(k) B(k-1) \overline{a} \overline{a}^{T} B(k-1)^{T} C(k)^{T} + D(k) R D(k)^{T}.$$
(20)

The FDI scheme in the following lemma will play a key role in deriving the main results of this paper. \Box

Lemma 4. With above derivation, we can decide whether the observations of system were lost and detect the lost information at instant k according to the following rule:

$$S_k > S_{th} \Longrightarrow \gamma_k = 0$$

$$S_k \le S_{th} \Longrightarrow \gamma_k = 1,$$
(21)

where

$$S_{th} = C(k) B(k-1) \bar{a} \bar{a}^{T} B(k-1)^{T} C(k)^{T} + D(k) R D(k)^{T}.$$
(22)

With the fault being detected, the missing information γ_k can be taken into consideration when designing the FDI-LMMSE filter. The fault observation can be replaced and isolated by $\hat{y}(k \mid k - 1)$. By the above approach, we can skip the error information at the instant *k* and use the correct information of pervious instant k - 1 to estimate the value of $\hat{x}(k \mid k)$ state at instant *k* directly.

Theorem 5. Consider the system represented by (8). Then the LMMSE $\hat{x}_{k|k}$ is given by

$$\widehat{x}_{k|k} = \sum_{i=1}^{N} \widehat{z}_i \left(k \mid k \right), \tag{23}$$

where $\hat{z}^{c}(k \mid k)$ satisfies the recursive equation

$$\widehat{z}_{\gamma_{k}}\left(k \mid k\right) = \widehat{z}\left(k \mid k-1\right) + \gamma_{k}Z\left(k \mid k-1\right)C(k)^{T}$$

$$\times \left[C(k)\widetilde{Z}(k \mid k-1)C(k)^{T} + D(k)RD(k)^{T}\right]^{-1}$$

$$\times \left(C\left(k\right)\widetilde{z}\left(k \mid k-1\right) + D\left(r_{k}\right)v\left(k\right)\right)$$
(24)

$$\widehat{z}(k+1 \mid k) = A(k)\,\widehat{z}(k \mid k) + B(k)\,\overline{a},\tag{25}$$

where $\hat{z}(0 \mid -1) = [\mu_0^1 x_0, \dots, \mu_0^N x_0]^T$.

Proof. Recall that observation estimator is given by (12).

Now, $\tilde{y}(k \mid k - 1)$ can be rewritten as the following equation:

$$\tilde{y}(k \mid k-1) = \gamma_k \left[C(k) \,\tilde{z}(k \mid k-1) + D(k) \,v(k) \right].$$
(26)

Considering the geometric argument as in [39], the estimator $\hat{z}(k \mid k-1)$ satisfies the following equations:

$$\hat{z} (k \mid k-1) = E\left(z (k) \left(y^{k-1}\right)^{T}\right) \operatorname{cov}\left(\left(y^{k-1}\right)^{-1} y^{k-1}\right) \quad (27)$$
$$\hat{z} (k \mid k) = \hat{z} (k \mid k-1) + E\left(\hat{z} (k) \, \tilde{y} (k \mid k-1)^{T}\right)$$
$$\times E\left(\tilde{y} (k \mid k-1) \, \tilde{y} (k \mid k-1)^{T}\right)^{-1} \qquad (28)$$
$$\times \left(y (k) - \hat{y} (k \mid k-1)\right).$$

From (26), we get that

$$E\left(\widehat{z}(k)\ \widetilde{y}(k\mid k-1)^{T}\right) = \gamma_{k}\left[\widetilde{Z}(k\mid k-1)\ C(k)^{T}\right].$$
 (29)

Because v_k is independent of $\{r_k, y(k-1)\}$, we have that

$$\left\langle \alpha^{T} D(r_{k}) v_{k}; \beta^{T} y(k-1) \right\rangle$$

= $E\left(\alpha^{T} D(r_{k}) v_{k} \beta^{T} y(k-1)\right)$
= $E\left(v_{k}^{T}\right) E\left[\alpha^{T} D(r_{k}) \beta^{T} y(k-1)\right]$
= 0, (30)

showing that $D(k)v_k$ is orthogonal to $\mathcal{L}(y^{k-1})$. Similar reasoning shows the orthogonality between $D(k)v_k$ and $\tilde{z}(k \mid k-1)$. Recalling that $\hat{z}(k \mid k-1) \in \mathcal{L}(y^{k-1})$ and $\tilde{z}(k \mid k-1)$ are orthogonal to $\mathcal{L}(y^{k-1})$, we can obtain that $\tilde{z}(k \mid k-1)$ is orthogonal to $\hat{z}(k \mid k-1)$. Then, from (27), the result can be obtained as follows:

$$\begin{aligned} \widehat{z}_{j}\left(k \mid k-1\right) &= E\left(z_{j}\left(k\right)\left(y^{k-1}\right)^{T}\right) \operatorname{cov}\left(\left(y^{k-1}\right)^{-1}y^{k-1}\right) \\ &= \sum_{i=1}^{N} \pi_{ij} A_{i} \widehat{z}_{i}\left(k-1 \mid k-1\right) + \sum_{i=1}^{N} \pi_{ij} \mu_{k-1}^{i} B_{i} a_{i}. \end{aligned}$$
(31)

From (11), (28) and (26), (29), we get that

$$\widehat{z}(k \mid k)$$

$$= \widehat{z}(k \mid k-1) + \gamma_k \widetilde{Z}(k \mid k-1) C(k)^T$$

$$\times \left[C(k)\widetilde{Z}(k \mid k-1)C(k)^T + D(k)RD(k)^T \right]^{-1}$$

$$\times \left(C(k)\widetilde{z}(k \mid k-1) + D(r_k)v(k) \right).$$
(32)

The positive-semidefinite matrices $\widetilde{Z}(k \mid k - 1)$ are obtained from

$$\widetilde{Z}(k \mid k-1) = Z(k) - \widehat{Z}(k \mid k-1).$$
(33)

And the recursive equation about $\widehat{Z}(k \mid k-1)$ is given as follows:

$$\begin{split} \widehat{Z}_{\gamma_{k}} \left(k \mid k \right) \\ &= \widehat{Z}_{\gamma_{k}} \left(k \mid k-1 \right) \\ &+ \gamma_{k}^{2} \widehat{Z}_{\gamma_{k}} \left(k \mid k-1 \right) C(k)^{T} \\ &\times \left(C(k) \widetilde{Z}_{\gamma_{k}} \left(k \mid k-1 \right) C(k)^{T} + D(k) R D(k)^{T} \right)^{-1} \\ &\times C \left(k \right) \widehat{Z}_{\gamma_{k}} \left(k \mid k-1 \right) \\ \widehat{Z}_{\gamma_{k}} \left(k \mid k-1 \right) \\ &= A \left(k-1 \right) \widehat{Z}_{\gamma_{k-1}} \left(k-1 \mid k-1 \right) A(k-1)^{T} \\ &+ B \left(k-1 \right) \overline{a} \overline{a}^{T} B(k-1)^{T}, \end{split}$$
(34)

where $\widehat{Z}(0 \mid -1) = z(0 \mid -1)z(0 \mid -1)^{T}$.

 $\widetilde{Z}(k+1 \mid k)$ can be derived directly as a recursive Riccati equation in the following derivation. In the following, we denote the linear operator

$$\Psi(\cdot,k): H^n \longrightarrow B\left(R^{Nn}\right) \tag{35}$$

by $\Gamma(k)$, in which $\Psi(\cdot, k)$ is

$$\Psi(\Gamma(k)) = \operatorname{diag}\left[\sum_{i=1}^{N} \pi_{ij} A_i Z_i(k) A_i^T\right]$$

$$-A(k) \left(\operatorname{diag}\left[Z_i(k)\right]\right) A(k)^T \ge 0.$$

$$\Box$$

Theorem 6. $\widetilde{Z}(k+1 \mid k)$ satisfies the following recursive Riccati equation:

$$\overline{Z}(k+1 \mid k) = A(k)\overline{Z}(k \mid k-1)A(k)^{T}$$

$$+ \Psi(\Gamma(k), k) + B(k)QB(k)^{T}$$

$$- \gamma_{k}^{2}A(k)\overline{Z}(k \mid k-1)C(k)^{T}$$

$$\times \left[C(k)\overline{Z}(k \mid k-1)C(k)^{T} + D(k)RD(k)^{T}\right]^{-1}$$

$$\times C(k)\overline{Z}(k \mid k-1)A(k)^{T},$$
(37)

where $\Gamma(k) = (Z_1(k), Z_2(k), \dots, Z_N(k))$ is given by the recursive equation (9) from Lemma 3.

Unlike the classical case, the sequence $\{\overline{Z}(k+1 \mid k)\}_{k \in \mathbb{Z}_+}$ is now random, which result from its dependence on the random sequence $\{\gamma_k\}_{k \in \mathbb{Z}_+}$.

Proof. Rewrite state equation in (8) as follows:

$$z (k + 1) = A (k) z (k) + M (k + 1) z (k) + B (k) \overline{a} + \vartheta (k),$$
(38)

where

$$M(k+1, j) = [m_1(k+1, j) \cdots m_N(k+1, j)],$$

$$m_i(k+1, j) = (1_{\{r_{k+1}=j\}} - \pi_{ij}) A_i 1_{\{r_k=i\}},$$

$$M(k+1) = \begin{bmatrix} M(k+1, 1) \\ \vdots \\ M(k+1, N) \end{bmatrix},$$

$$\vartheta(k) = \begin{bmatrix} 1_{\{r_{k+1}=1\}} B_1 w(k) \\ \vdots \\ 1_{\{r_{k+1}=N\}} B_N w(k) \end{bmatrix}.$$
(39)

From (32), we define

$$T(k) = -A(k)\widetilde{Z}(k | k-1)C(k)^{T} \times \left[C(k)\widetilde{Z}(k | k-1)C(k)^{T} + D(k)RD(k)^{T}\right]^{-1}.$$
(40)

From (25) and (32), we have that

$$\begin{aligned} \widehat{z}_{\gamma_{k}}\left(k+1\mid k\right) &= A\left(k\right)\widehat{z}\left(k\mid k-1\right) \\ &+ \gamma_{k}T\left(k\right)C\left(k\right)\widetilde{z}\left(k\mid k-1\right) \\ &+ B\left(k\right)\overline{a} + \gamma_{k}T\left(k\right)D\left(k\right)w_{k}. \end{aligned}$$
(41)

Then from (41) and (38), we get that

Therefore, at this point, we obtain the recursive equation for $\widetilde{Z}(k \mid k - 1)$ as follows:

$$\begin{split} \widetilde{Z}_{\gamma_{k}}\left(k+1\mid k\right) &= \left(A\left(k\right)+\gamma_{k}T\left(k\right)C\left(k\right)\right) \\ &\times \widetilde{Z}\left(k\mid k-1\right)\left(A(k)+\gamma_{k}T(k)C(k)\right)^{T} \\ &+ E\left(M\left(k+1\right)z\left(k\right)z(k)^{T}M(k+1)^{T}\right) \\ &+ E\left(\vartheta\left(k\right)\vartheta(k)^{T}\right) \\ &+ \gamma_{k}^{2}T\left(k\right)D\left(k\right)RD(k)^{T}T(k)^{T}. \end{split}$$

$$\end{split}$$

$$(43)$$

$$E\left(M\left(k+1\right)z\left(k\right)z\left(k\right)^{T}M\left(k+1\right)^{T}\right) = \Psi\left(\Gamma\left(k\right),k\right)$$

$$E\left(\vartheta\left(k\right)\vartheta\left(k\right)^{T}\right) = B\left(k\right)QB\left(k\right)^{T}.$$
(44)

Substituting (44) into (43) yields the recursive equation for $\widetilde{Z}(k \mid k - 1)$ as

$$\overline{Z}_{\gamma_{k}}(k+1 \mid k) = (A(k) + \gamma_{k}T(k)C(k))\overline{Z}(k \mid k-1) \\
\times (A(k) + \gamma_{k}T(k)C(k))^{T} \\
+ \Psi(\Gamma(k), k) + B(k)QB(k)^{T} \\
+ \gamma_{k}^{2}T(k)D(k)RD(k)^{T}T(k)^{T}.$$
(45)

4. Stability of the State Estimator

As we all see, the intermittent observations are the source of potential instability. From Theorem 6, however, the error covariance matrix obtained from the LMMSE can be rewritten in terms of a recursive Riccati equation of γ_k . In this section, based on that following assumptions hold, we show that the proposed estimator is stable as provided in our paper.

Assumption 7. $\{r_k, k = 0, 1 \dots\}$ is assumed to be ergodic Markov chain.

Assumption 8. System (1) is mean square stable (MSS) according to the definition in [35].

First, (37) describes a recursive Riccati equation for $\widetilde{Z}(k+1 \mid k)$. We should establish now its convergence when $k \rightarrow \infty$. It follows from Assumption 2 that $\lim_{k\to\infty} P(r_k = i)$ exists and it is independent of r_0 . We define

$$\mu_i = \lim_{k \to \infty} P\left(r_k = i\right) = \lim_{k \to \infty} \mu_i\left(k\right). \tag{46}$$

We redefine the matrix as follows:

$$A \triangleq \begin{pmatrix} \pi_{11}A_1 & \cdots & \pi_{1N}A_N \\ \vdots & \ddots & \vdots \\ \pi_{N1}A_1 & \cdots & \pi_{NN}A_N \end{pmatrix},$$

$$B \triangleq \operatorname{diag} \left[\left[\left(\pi_{1j}\mu_1 \right)^{1/2} B_1 \cdots \left(\pi_{Nj}\mu_N^N \right)^{1/2} B_N \right] \right],$$

$$C \triangleq \left[C_1, C_2, \dots, C_N \right], \qquad D \triangleq \left[D_1 \mu_1^{1/2} \cdots D_N \mu_N^{1/2} \right].$$
(47)

Then, we give the following facts and lemmas for system, which will be used in the proof of stability of the covariance matrix of estimation error.

With regard to Assumptions 2 and 7 and Proposition 3.36 in [35], $\Gamma(k) \rightarrow \Gamma$ as $k \rightarrow \infty$, where $\Gamma = \{Z_1, Z_2, \dots, Z_N\}$ is the unique solution that satisfies

$$Z_{j} = \sum_{i=1}^{N} \pi_{ij} \left(A_{i} Z_{i} A_{i}^{T} + \mu_{i} B_{i} \left(a_{i} a_{i}^{T} + Q \right) B_{i}^{T} \right).$$
(48)

Then we have $\inf_{l\geq k}\mu_i(l) > 0$ holding for all $i \in \mathbf{N}$ (since $\exists l$, we have $\mu_i(l) \to \mu_i > 0$ as $k \to \infty$). Defining $\alpha_i(k) = \inf_{l\geq k}\mu_i(k+l)$, then we get

$$\mu_i(k+\kappa) \ge \alpha_i(k) \ge \alpha_i(k-1), \quad k = 1, 2, ...; \ i \in \mathbb{N}.$$
 (49)

At the same time, $\alpha_i(k) \rightarrow \mu_i$ $(k \rightarrow \infty)$ exponentially fast.

From (37), as $k \rightarrow \infty$, we obtain the mean state covariance as follows:

$$\overline{\widetilde{Z}}(k+1 \mid k) = \lim_{k \to \infty} \widetilde{Z}(k+1 \mid k)$$

$$= \lim_{k \to \infty} E\left[\widetilde{Z}(k+1 \mid k)\right]$$

$$= A\overline{\widetilde{Z}}(k \mid k-1) A^{T} + \Psi(\Gamma(k))$$

$$+ B(k) QB(k)^{T} - \eta \overline{\widetilde{Z}}(k \mid k-1) C^{T}$$

$$\times \left[C\overline{\widetilde{Z}}(k \mid k-1)C^{T} + D(k)RD(k)^{T}\right]^{-1}$$

$$\times C\overline{\widetilde{Z}}(k \mid k-1) A^{T}.$$
(50)

An operator is introduced for any positive-semidefinite matrix *X* as follows:

$$\mathcal{T} = -AXC^{T} \left(CXC^{T} + DRD^{T} \right)^{-1}.$$
 (51)

Then

$$\mathcal{T}(k) = -A\overline{\widetilde{Z}}(k \mid k-1) C^{T} \left(C\overline{\widetilde{Z}}(k \mid k-1)C^{T} + DRD^{T} \right)^{-1}.$$
(52)

Define now $\overline{\Gamma}(k) = (\overline{Z}_1(k), \dots, \overline{Z}_N(k)) \in H^{n+}$ with $\overline{Z}_i(0) = 0, j \in \mathbf{N}^+$, and

$$\bar{Z}_{j}(k+1) = \sum_{i=1}^{N} \pi_{ij} \left(A_{i} \bar{Z}_{i}(k) A^{T} + \alpha_{i}(k) B_{i} Q B_{i}^{T} \right).$$
(53)

Lemma 9 (see [35]). $\overline{\Gamma}(k) \xrightarrow{k \to \infty} \Gamma$ and for each $k = 0, 1, 2, \dots$, one can get that

$$\Gamma\left(k+\kappa\right) \ge \bar{\Gamma}\left(k\right) \ge \bar{\Gamma}\left(k-1\right). \tag{54}$$

Now, one defines

$$\Upsilon (k+1) = A \Upsilon (k) A^{T} + \operatorname{diag} \left[\sum_{i=1}^{N} \alpha_{i} (k) \pi_{ij} B_{i} Q B_{i}^{T} \right] + \Psi \left(\bar{\Gamma} \right)$$
$$- \eta A \Upsilon (k) C^{T} \left(C \Upsilon (k) C^{T} + \bar{D} (k) R \bar{D} (k)^{T} \right)^{-1}$$
$$\times C P_{1} (k) A^{T},$$
(55)

where $\Upsilon(0) = 0$, $\overline{D}(k) = [D_1\alpha_1(k)^{1/2} \cdots D_N\alpha_N(k)^{1/2}]$. From the definition of κ and condition of $DRD^T > 0$, one notices that the inverse of $C\Upsilon(k)C^T + DRD^T$ exists. **Lemma 10.** For each $k = 0, 1 \cdots$, one gets that

$$\Upsilon(k) \le \Upsilon(k+1) \le \widetilde{Z}(k+1+\kappa \mid k+\kappa).$$
(56)

Proof. In order to deduce (56), we define

$$\mathcal{M}(k) = -A\Upsilon(k) C^{T} (C\Upsilon(k)C^{T} + \bar{D}(k)R\bar{D}(k)^{T})^{-1}.$$
 (57)
Then, if $\Upsilon(k) \leq \overline{\widetilde{Z}}(k + \kappa \mid k + \kappa - 1),$
 $\Upsilon(k + 1)$

$$= \left(A + \sqrt{\eta}\mathcal{T}(k+\kappa)C\right)\Upsilon(k)\left(A + \sqrt{\eta}\mathcal{T}(k+\kappa)C\right)^{T} + \Psi\left(\bar{\Gamma}(k)\right) + \eta\mathcal{T}(k+\kappa)\bar{D}(k)R\bar{D}(k)\mathcal{T}(k+\kappa)^{T} - \eta\left(\mathcal{T}(k+\kappa) - \mathcal{M}(k)\right)\left(C\Upsilon(k)C^{T} + \bar{D}(k)R\bar{D}(k)\right) \times\left(\mathcal{T}(k+\kappa) - \mathcal{M}(k)\right)^{T} + \operatorname{diag}\left[\sum_{i=1}^{N}\alpha_{i}\left(k\right)\pi_{ij}B_{i}QB_{i}^{T}\right] \leq \left(A + \sqrt{\eta}\mathcal{T}(k+\kappa)C\right)\overline{\widetilde{Z}}\left(k+\kappa \mid k+\kappa-1\right) \times\left(A + \sqrt{\eta}\mathcal{T}(k+\kappa)C\right)^{T} + \operatorname{diag}\left[\sum_{i=1}^{N}\mu_{i}\left(k+\kappa\right)\pi_{ij}B_{i}QB_{i}^{T}\right] + \Psi\left(\bar{\Gamma}(k+\kappa)\right) + \eta\mathcal{T}(k+\kappa)D\left(k+\kappa\right)RD\left(k+\kappa\right)\mathcal{T}(k+\kappa)^{T} = \overline{\widetilde{Z}}\left(k+1+\kappa \mid k+\kappa\right).$$
(58)

Obviously, when $\Upsilon(0) = 0 \leq \overline{\widetilde{Z}}(\kappa | \kappa - 1)$, it yields $\Upsilon(k) \leq \overline{\widetilde{Z}}(k+\kappa | k+\kappa-1), k = 0, 1, 2 \cdots$. Similarly if $\Upsilon(k-1) \leq \Upsilon(k)$, based on (49) and (54), we have

$$\Upsilon(k)$$

$$= (A + \sqrt{\eta}\mathcal{M}(k)C) \Upsilon(k-1) (A + \sqrt{\eta}\mathcal{M}(k)C)^{T} + \Psi(\bar{\Gamma}(k)) + \eta\mathcal{M}(k)\bar{D}(k)R\bar{D}(k)\mathcal{M}(k)^{T} + \eta(\mathcal{M}(k) - \mathcal{M}(k-1)) \times (C\Upsilon(k)C^{T} + \bar{D}(k)R\bar{D}(k))(\mathcal{M}(k) - \mathcal{M}(k-1))^{T} + diag\left[\sum_{i=1}^{N} \alpha_{i}(k-1)\pi_{ij}B_{i}QB_{i}^{T}\right] \leq (A + \sqrt{\eta}\mathcal{M}(k)C)\Upsilon(k)(A + \sqrt{\eta}\mathcal{M}(k)C)^{T} + \Psi(\bar{\Gamma}(k)) + diag\left[\sum_{i=1}^{N} \alpha_{i}(k)\pi_{ij}B_{i}QB_{i}^{T}\right] + \eta\mathcal{M}(k)\bar{D}(k)R\bar{D}(k)\mathcal{M}(k)^{T} = \Upsilon(k+1).$$
(70)

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Since $\Upsilon(0) = 0 \le \Upsilon(1)$, the induction argument is completed for $\Upsilon(k) \le \Upsilon(k+1)$.

Theorem 11. Suppose that Assumptions 7 and 8 hold. Consider that the algebraic Riccati equation

$$P = APA^{T} + \Psi(\Gamma) + BQB^{T} - \eta APC [CPC^{T} + DRD]^{-1}CPA^{T}$$
(60)

satisfies (48), where $\Gamma = \{Z_1, Z_2, ..., Z_N\}$. Then, there exists a unique nonnegative definite solution P to (60). $r_{\sigma}(A) \leq 1$, and for any $\Gamma(0) = \{Z_1(0), ..., Z_N(0)\}, Z_i(0) \geq 0, i = 1 \cdots N$, and $\overline{\widetilde{Z}}(0 \mid -1) = \widetilde{Z}(0 \mid -1) \geq 0$, one has $\overline{\widetilde{Z}}(k+1 \mid k)$ given by (50) satisfying $\overline{\widetilde{Z}}(k+1 \mid k) \rightarrow P$, $k \rightarrow \infty$.

Proof. Due to MSS of 5.38 [35], we have from Proposition 3.6 in chapter 3 [35] that $r_{\sigma}(A) < 1$. According to the standard results for algebraic Riccati equation there is a unique positive-semidefinite solution $P \in B(\mathbb{R}^{Nn})$ to (60). And moreover $r_{\sigma}(A + \sqrt{\eta}\mathcal{T}(P)C) < 1$.

From Theorem 11, we get that P satisfied

$$P = (A + \sqrt{\eta}\mathcal{T}(P)C)P(A + \sqrt{\eta}\mathcal{T}(P)C)^{T} + \Psi(\Gamma) + BQB^{T} + \eta\mathcal{T}(P)DRD^{T}\mathcal{T}(P)^{T}.$$
(61)

Define $P(0) = \overline{\widetilde{Z}}(0 \mid -1) = \widetilde{Z}(0 \mid -1)$ and

$$P(k+1) = (A + \sqrt{\eta}\mathcal{T}(P)C)P(k)(A + \sqrt{\eta}\mathcal{T}(P)C)^{T} + \Psi(\Gamma(k)) + B(k)QB(k)^{T}$$
(62)
+ $\eta\mathcal{T}(P)D(k)RD(k)^{T}\mathcal{T}(P)^{T}.$

Then (50) can be rewritten as

$$\overline{\overline{Z}}(k+1 | k)$$

$$= (A + \sqrt{\eta} \mathcal{T}(P) C) \overline{\overline{Z}}(k | k-1) (A + \sqrt{\eta} \mathcal{T}(P) C)^{T}$$

$$+ \Psi(\Gamma(k)) + B(k) QB(k)^{T}$$

$$+ \eta \mathcal{T}(P) D(k) RD(k) \mathcal{T}(P)^{T} - \eta (\mathcal{T}(k) - \mathcal{T}(P))$$

$$\times \left[C\overline{\overline{Z}}(k | k-1) C^{T} + D(k) RD(k)^{T} \right]$$

$$\times (\mathcal{T}(k) - \mathcal{T}(P))^{T}.$$
(63)

(59)

Suppose that $P(k) \ge \overline{\widetilde{Z}}(k \mid k-1)$, we have that

$$P(k+1) - \overline{\widetilde{Z}}(k+1 | k)$$

$$= (A + \sqrt{\eta} \mathcal{F}(P) C)$$

$$\times (P(k) - \overline{\widetilde{Z}}(k | k-1)) (A + \sqrt{\eta} \mathcal{F}(P) C)^{T}$$

$$+ \eta (\mathcal{F}(k) - \mathcal{F}(P))$$

$$\times (C\overline{\widetilde{Z}}(k | k-1) C^{T} + D(k) RD(k)^{T})$$

$$\times (\mathcal{F}(k) - \mathcal{F}(P))^{T}.$$
(64)

By definition, $P(0) = \widetilde{Z}(0 \mid 1)$. Suppose that $P(k) \ge \widetilde{Z}(k \mid k - 1)$. From (64), we have that $P(k + 1) \ge \overline{\widetilde{Z}}(k + 1 \mid k)$. Therefore we have shown by induction that $P(k) \ge \overline{\widetilde{Z}}(k \mid k - 1)$ for all $k = 0, 1, 2 \cdots$. From MSS and ergodicity of the Markov chain we have that $\Gamma(k) \xrightarrow{k \to \infty} \Gamma, D(k) \xrightarrow{k \to \infty} D$, and $B(k) \xrightarrow{k \to \infty} B$ exponentially fast. From $r_{\sigma}(A + \sqrt{\eta} \widetilde{\mathcal{T}}(P)C) < 1$ and same reasoning as in the proof of proposition 3.36 in [35] we have that $P(k) \to P$ as $k \to \infty$, where *P* satisfies

$$\bar{P} = \left(A + \sqrt{\eta}\mathcal{F}(P)C\right)\bar{P}\left(A + \sqrt{\eta}\mathcal{F}(P)C\right)^{T} + \Psi\left(\Gamma\right) + BQB^{T} + \eta\mathcal{F}(P)DRD^{T}\mathcal{F}(P)^{T}.$$
(65)

And *P* is the unique solution to (65). Recalling that *P* satisfies (62), we get that *P* is also a solution to (65) and from uniqueness, $\bar{P} = P$. Then, we obtain that

$$\overline{\widetilde{Z}}(k \mid k-1) \le P. \tag{66}$$

And $P(k) \rightarrow P$. From (66) and (56) in Lemma 10 it follows that $0 \leq \Upsilon(k) \leq \Upsilon(k+1) \leq P(k+1+\kappa)$. And thus we can conclude that $\Upsilon(k) \rightarrow \Upsilon$ whenever $k \rightarrow \infty$ for some $\Upsilon \geq 0$. Moreover, from the fact that $\alpha_i(k) \xrightarrow{k \rightarrow \infty} \mu_i$ and $\overline{\Gamma}(k) \xrightarrow{k \rightarrow \infty} \Gamma$, we have that Υ satisfies (60).

From uniqueness of the positive-semidefinite solution to (60), we can conclude that $\Upsilon = P$. From (66) and (56), $\Upsilon(k) \leq \overline{\widetilde{Z}}(k + \kappa \mid k + \kappa - 1) \leq P(k + \kappa)$ and since $\Upsilon(k) \rightarrow P$ and $P(k) \rightarrow P$ as $k \rightarrow \infty$, we get that $\overline{\widetilde{Z}}(k \mid k - 1) \xrightarrow{k \rightarrow \infty} P$. \Box

The upper bound P for the error covariance matrix to a stationary value for linear minimum mean square error (LMMSE) estimation can be easily obtained. It is described that if the system is MSS and the missing information is detected, then the error covariance matrix will converge to the unique nonnegative definite solution of an algebraic Riccati equation associated with the problem.

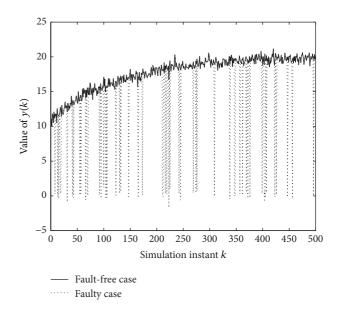


FIGURE 1: Value of observations.

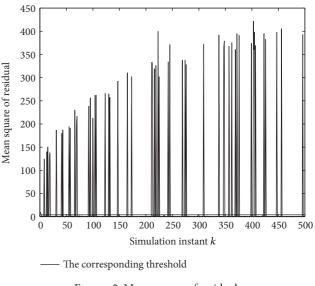
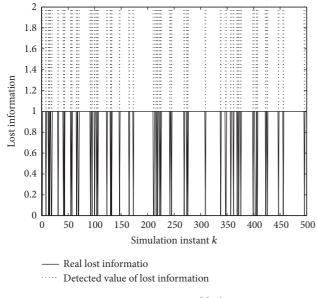


FIGURE 2: Mean square of residual.

5. Numerical Example

In order to evaluate the performance of our method, in this section, we are going to use a scalar MJLS described by the following equations:

$$\begin{aligned} x_{k+1} &= A_{r_k} x_k + B_{r_k} \left(a_{r_k} + w_k \right) \\ y_k &= \gamma_k C_{r_k} x_k + D_{r_k} v_k \\ A_1 &= \begin{pmatrix} 1 & 0.995 \\ 0 & 1 \end{pmatrix}, \qquad A_2 &= \begin{pmatrix} 1 & 0.99 \\ 0 & 1 \end{pmatrix}. \end{aligned}$$



20 19 18 17 16 Value of x15 14 17 13 16.95 12 16.9 11 144 146 148 10 50 100 150 200 250 300 350 400 450 500 0 Simulation instant k --- Fault-free case ····· Faulty case True x

FIGURE 3: Detection of fault.

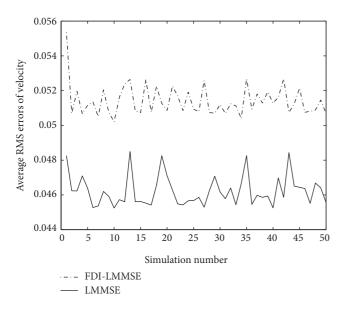


FIGURE 4: Average RMS target position error.

$$B_{1} = B_{2} = \begin{pmatrix} 0.1\\0 \end{pmatrix}, \qquad C_{1} = C_{2} = \begin{pmatrix} 1\\0 \end{pmatrix},$$
$$D_{1} = D_{2} = \begin{pmatrix} 5\\0 \end{pmatrix}, \qquad a_{1} = 1, \qquad a_{2} = 2,$$
(67)

where $x_k(1, 1)$, $x_k(2, 1)$, and a_k denote the target position, velocity, and acceleration, respectively. The initial state x_0 is normally distributed with mean 10 and variance 1. $r_k \in \{1, 2\}$, and w_k , v_k are independent white noise sequences with

FIGURE 5: Comparison between state estimators of x(1) under no faulty case and faulty case, respectively, and real state value.

covariance of 0.1², and $\mu_0^1 = \mu_0^2 = 0.5$. The transition probability matrix for the finite-state Markov chain is

$$\Pi = \begin{pmatrix} 0.6 & 0.4 \\ 0.4 & 0.6 \end{pmatrix},$$

$$(\gamma_k = 1) = 0.9, \qquad P(\gamma_k = 0) = 0.1.$$
(68)

To assess the performance of algorithms, the average root mean square (RMS) error based on H times Monte-Carlo simulation is defined as

RMS =
$$\frac{1}{H} \frac{1}{T} \sum_{i=1}^{H} \sum_{k=1}^{T} \left[\left(x_k^i - \hat{x}_k^i \right)^2 \right]^{1/2},$$
 (69)

where the time step T is chosen as 500, H = 50.

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The simulation results are obtained as follows. Figure 5 presents the real states and their estimators subject to faultfree case and faulty case, respectively, based on the given path. Figure 1 shows the observations with lost data from unreliable channel and observations from reliable channel. As the proposed algorithm can be thought of as a generalization of the well-known LMMSE filtering, we denote it by FDI-LMMSE filtering in the simulation. The RMS in the position of FDI-LMMSE filtering in the faulty case is compared with that of LMMSE filtering in the fault-free case in the Figure 4. It can be shown in Figures 2 and 3 that the residual can deliver fault alarms soon after the fault occurs. From the simulation results, we can see that the obtained linear estimator for systems with random missing data are tracking well to the real state value, which is the estimation scheme proposed in this paper produces good performance.

6. Conclusions

This paper has addressed the estimation problem for MJLSs with random missing data. Random missing data introduced by the network is modeled as Bernoulli distribution variable. By usage of an observer-based FDI as a residual generator, the design of FDI-LMMSE filter has been formulated in the framework of LMMSE filtering. Complete analytical solution has been obtained by solving the recursive Riccati equations. It has been proved from theorem derivation and a numerical example simulation that the proposed state estimator is effective.

Conflict of Interests

The authors declare that there is no conflict of interests regarding the publication of this paper.

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